## 16.4 Energy-Autonomous Fever Alarm Armband Integrating Fully Flexible Solar Cells, Piezoelectric Speaker, Temperature Detector, and 12V Organic Complementary FET Circuits

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Three key requirements for wearable healthcare and biomedical devices are the mechanical flexibility, the wireless interface, and the energy autonomy, because unobtrusive and maintenance-free devices are needed for the constant monitoring of vital human health data. Previously reported flexible healthcare and biomedical devices, however, requires wired connection [1,2] or wireless power transmission [3,4]. A flexible energy autonomus healthcare device with a wireless interface, a fever alarm armband (FAA) integrating fully flexible solar cells, a piezoelectric speaker, a temperature detector, and 12V organic complementary FET (CFET) circuits is presented here. The system is also noteworthy for sound generation, with organic circuits driving the speaker.

Figure 16.4.1 shows a photograph of FAA with the size of  $30x18cm^2$  and its use case. All components in FAA are mechanically flexible. FAA is looped around an upper arm of a patient in a hospital room, and the temperature detector monitors the underarm temperature of the patient. A 220µm thick amorphous silicon solar cells (PowerFilm P3-37) attached outside of the upper arm generate the power (e.g. 3mW at 1500lx). When high fever is detected, a 52µm thick piezoelectric speaker with polyvinylidene difluoride (PVDF) makes a sound to alarm a nurse. Organic CFET circuits and the temperature detector are fabricated on a 50µm thick flexible polyimide film, and the solar cells and the speaker are attached on it.

Figure 16.4.2 shows a circuit schematic of FAA. An active voltage limiter (AVL) regulates the power supply voltage ( $V_{DD}$ ) of the FAA, because  $V_{DD}$  changes depending on both the illuminance for the solar cells and the power consumption of the FAA. When the measured temperature is higher than the preset threshold temperature (T<sub>t</sub>), the ring oscillator (RO) starts oscillation and the speaker makes a sound. Two design challenges of the FAA are as follows. 1) In the temperature detector, variable  $T_t$  by users is required, because the needed  $T_t$  depends on the patient. 2) The voltage regulation of  $V_{DD}$  is required, because the output voltage change of the solar cells due to the variation of the indoor lighting is large (e.g. 10V at 500lx and 20V at 3000lx) and the overvoltage breaks down the organic transistors. In the organic electronics, however, the voltage regulation is difficult, because a voltage reference and Zener diodes are not available. To solve these problems, the following two solutions are proposed in this paper. 1) A voltage controlled threshold temperature tuning (VCTTT) enables the tuning of T<sub>t</sub>. 2) AVL is the world's first voltage regulation in the organic electronics.

Figures 16.4.3 (a) and (b) show a circuit schematic and the measured temperature dependence of the temperature detector, respectively. When the temperature is increased, the resistance of a resistive-type flexible thermal sensor increases by more than 10<sup>4</sup> times at T<sub>t</sub>, and EN changes from high to low. In the proposed VCTTT, the gate voltage (V<sub>G</sub>) of p-type transistor operating in the sub-threshold region is varied to tune T<sub>t</sub>. When V<sub>G</sub> is varied from 10V to 12V, T<sub>t</sub> changes from 36.5°C to 38.5°C, which indicates 1°C/V tunability. The control of V<sub>G</sub> could be achieved by selectively cutting the wires of a resistor ladder, though a V<sub>G</sub> generator is not implemented in the FAA.

In the FAA, a careful choice of the frequency of the sound is required, because the frequency is determined by both RO and the speaker. Figure 16.4.4(a) shows the measured  $V_{DD}$  dependence of the RO frequency. The maximum frequency of RO is 6kHz, because the maximum  $V_{DD}$  ( $V_{MAX}$ ) is 15V due to the breakdown voltage of the transistors. In contrast, the speaker does not make a sound below 1.8kHz. Therefore, the target frequency of the sound is selected to be 4.7kHz, because the speaker itself has the maximum sound pressure level (SPL) at 5kHz within a range of 1.8kHz to 6kHz. The corresponding target  $V_{DD}$  for RO is 12V. Figure 16.4.4 (a) also shows  $V_{DD}$  dependence of SPL of the speaker at RO

frequency measured with a sound level meter (RION NL-32) at 30cm distance. The minimum  $V_{\text{DD}}$  ( $V_{\text{MIN}}$ ) is defined as 9V at SPL of 10dB corresponding to the background noise. Figure 16.4.4(b) shows the measured waveform of the input signal of the speaker ("Out" in Fig. 16.4.2) at 12V and 4.7kHz. The power consumption of RO and the driver is 2.2mW. Figures 16.4.4 (c) and (d) show the measured spectra of SPL without and with the speaker, respectively. Figure 16.4.4(c) shows background noise in the measurement room. As shown in Fig. 16.4.4(d), the speaker successfully makes a sound with SPL of 28dB at 4.7kHz and the SNR is 18dB.

Figures 16.4.5(a) and (b) show circuit schematics of the proposed AVL and a conventional diode clamp for comparison, respectively. AVL is a kind of a shunt regulator, and is composed of a constant voltage generator, two CFET inverters. and a power transistor (M5). The constant voltage  $(V_1)$  and a logical threshold voltage of the inverter  $(-V_{DD}/2)$  are compared, and the output of the inverter  $(V_2)$ is varied to regulate  $V_{\mbox{\tiny DD}}.$  The essence of the constant voltage generator is stacked diode-connected transistor (M1) with small W/L and OFF transistor (M2) with large W/L. When the channel length modulation effect of the transistors is neglected,  $V_3$  is constant without a dependence on  $V_{\text{DD}}$ , because M2 is a constant current source with the OFF current and  $V_{\text{DS}}$  (=V\_{\text{GS}}) of M1 is determined by the OFF current. To increase the constant voltage  $V_3$  to  $V_{DD}/2$  (=6V), M3 and M4 are added, and  $V_1=2xV_3$ . Figure 16.4.5(c) shows the measured  $V_{DD}$  dependence of  $V_1$ and  $V_2$  in AVL. When  $V_{DD}$  is within 7V to 15V,  $V_1$  stays around 6.5V and  $\Delta V_1/\Delta V_{DD}$ =6% due to the channel length modulation. Figure 16.4.5(d) shows the measured  $V_{DD}$  dependence of  $I_D$  of AVL and the conventional diode clamp. In Fig. 16.4.5(d), the maximum  $\Delta I_D/\Delta V_{DD}$  of the diode clamp and AVL is 63µA/V and 121 $\mu$ A/V, respectively, which indicates that AVL increases  $\Delta I_D / \Delta V_{DD}$  by 1.9 times.

The impact of AVL on the operational illuminance range of FAA is shown. Figures 16.4.6(a), (b) and (c) show the measured illuminance dependence of V<sub>DD</sub> without the voltage limiter, with the conventional diode clamp, and with the proposed AVL, respectively. Two lines in each figure show with and without activating the load circuits (RO and the driver) in Fig. 16.4.2. The maximum and minimum operational illuminance are determined by V<sub>MAX</sub> (=15V) and V<sub>MIN</sub> (=9V), respectively. Without the voltage limiter in Fig. 16.4.6(a), the operational illuminance range is 300lx. By adding the conventional diode clamp in Fig. 16.4.6(b), the range is 1900lx. In the proposed AVL in Fig. 16.4.6(c), the range is 2200lx. AVL simultaneously achieves the minimum illuminance of Fig. 16.4.6(a) and the maximum illuminance of Fig. 16.4.6(b) thanks to the sharp I<sub>D</sub> control shown in Fig. 16.4.5(d). Thus, AVL increases the operational illuminance range from 300lx to 2200lx by 7.3 times.

Figure 16.4.7 shows micrographs of the organic circuits and a summary of the organic CFET. The organic semiconductors for p-type and n-type transistors are DNTT [4] and PDI-8CN2 [5], respectively. The minimum gate length is 10µm.

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Figure 16.4.5: (a) Proposed active voltage limiter (AVL). (b) Conventional diode clamp. (c) Measured  $V_{DD}$  dependence of  $V_1$  and  $V_2$  in AVL. (d) Measured  $V_{DD}$  dependence of  $I_D$ .



Figure 16.4.6: Measured illuminance dependence of  $V_{DD}$  (a) without voltage limiter, (b) with conventional diode clamp, and (c) with proposed AVL.

